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Improve Gain, Noise Figure And Spurious Free Dynamic Range In Intensity Modulation Analog Photonic Link Using Digital Signal Postprocessing Compensation By Linearized Downconverting Algorithm.

Kamleshshewar[#1] ,Namrata Sahayam[#2]

[#1]Researcher, Department of Electronics & Communication Engineering, Jabalpur Engineering College, Jabalpur,(M.P.) ,India

[#2]Professor, Department of Electronics & Communication Engineering, Jabalpur Engineering College, Jabalpur, (M.P.),India

A B S T R A C T

In this paper, we describe technique for performance improvement in analog photonics link incorporating digital signal post processing compensation technique by linearized down converting algorithm. We are investigating the performance components in this link and discuss the extent to which their performance varies with frequency, and show which of these components frequency-dependent parameters which affect; figure of merit in terms of gain, noise figure, and spurious free dynamic range. The proposed linearization does not require knowing the precise transfer function of the whole nonlinear system, which is achieved by directly acquiring the output third-order intercept point (IMD3) from Down converting algorithm. Using a high-performance improved link noise figure and gain are obtained while the nonlinearity can be also be compensated by the proposed algorithm. In our previous work measured noise figure and gainof the photonic link are 8.9 and 27.5 dB, respectively and spurious-free dynamic range of 128.3 dB in 1-Hz bandwidth which when compared by using linearized down converting algorithm improves to give noise figure and gain of the photonics link as 2.7 and 31.08 dB and spurious-free dynamic range of 130.49dB in 1-Hz bandwidth for a digitizer noise limited analog photonics link without using any cascade structure of pre and post amplifier

Keywords- Fiber optics, Radio frequency photonics, Digital signal post processing compensation technique, Low biasing.

1. Introduction

Analog photonics is the study of photonic devices, such as lasers and photo detectors, performing operations at microwave frequencies, and the application of these devices to microwave systems. The impact of photonics on digital communication systems is extensive and well known. Fiber optics carry massive amounts of data between users and services around the globe. These systems are finding applications in shorter and shorter distances, from long-distance telecommunications, to communication between servers in data centres, to interconnects within computers themselves.

Analog photonic link has become very attractive during the past years for both commercial and military applications, such as antenna remoting, radio astronomy, etc. In some challenging applications such as military use, analog photonic link must meet the stringent performance requirements in terms of dynamic range, gain, and other figures of merit.[1] Spurious free dynamic range (SFDR) is an important figure of merit that is usually used to evaluate the degree of linearity in analog photonics link . The inherent nonlinear characteristics of Mach-Zehnder modulator (MZM) introduces both harmonic product and inter-modulation distortions of the analog RF signal, of which the third-order intermodulation distortions (IMD3) are most pronounced and limit the SFDR of the system .[2] the third-order intermodulation distortion (IMD3) components should be particularly considered as they lie very close to the radio frequency (RF) carriers and are impossible to be eliminated by simple RF filtering. The IMD3 will then significantly reduce the SFDR performance of the link [3].

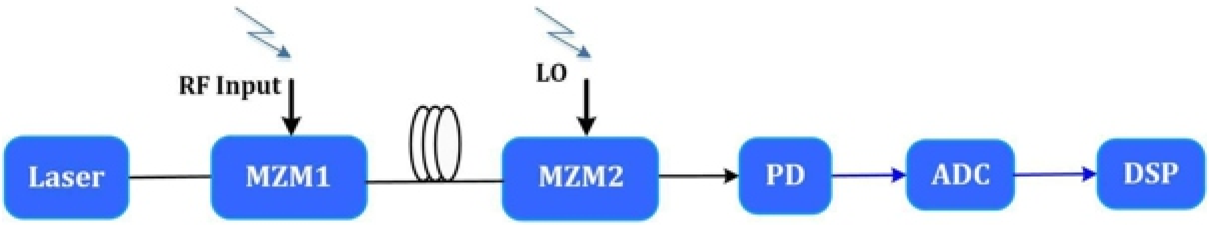

Fig. 1.General IMDD analog photonic link with optical down-conversion and DSP-based linearization. LO: local oscillator; ADC: analog-to-digital converter[1]

Multiple link parameters related to the laser, the modulator, and thephotodiode should be precisely known, which may otherwise result in significant imperfections. Furthermore, the SFDR capability does not tell the complete story for an analog photonics link, the gain and noise figure (NF) are also very important metrics that generally used to quantify the performance.[4] The design of the analog photonics systems requires careful consideration of all the performance parameters in terms of SFDR, gain and NF [6].

Downconversion Theory

Figure2. is a simplified block diagram of the down conversion scheme .A continuous-wave (CW) laser is modulated by MZM and transmitted over fiber to the receiver. In the receiver, the signal enters a second MZM that is driven by a strong microwave LO.Local Oscillator (LO) and the amplified two-tone input RF signal is converting to IF. This process is known as down conversion.

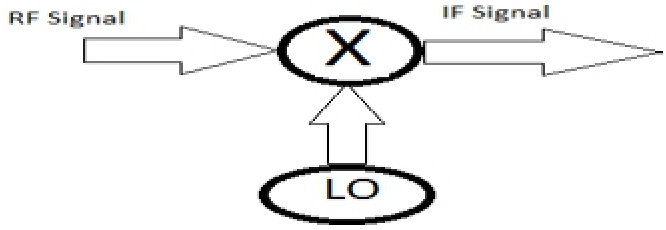

Fig. 2 Down-converted electrical signal

The optical down-conversion combined with DSP has been considered as a promising strategy forremote RF receiving[7]. Digital linearization for phase and polarization modulation has been widely discussed, for its linear modulation and demodulation at both ends[8].

Mach-Zehnder Modulator

Optical modulators are used for electrically controlling the output amplitude or the phase of the light wave passing through the device. To reduce the device size and the driving voltage, waveguide-based modulators are used for communication applications. Mach-Zehnder Modulator is use for the power of the input is split equally into the two output waveguides of the first directional coupler.

Our proposed method of the analog photonics systems requires careful consideration of all the improve parameters in terms of SFDR, gain and NF .

II. Principle of the Scheme

Analog photonic link using digital signal post processing compensation by linearized down converting algorithm without any cascade structure of pre and post amplifier .The schematic architecture of the proposed method shown in Fig. 3(a).and The block diagram of the proposed method shown in Fig.3(b).

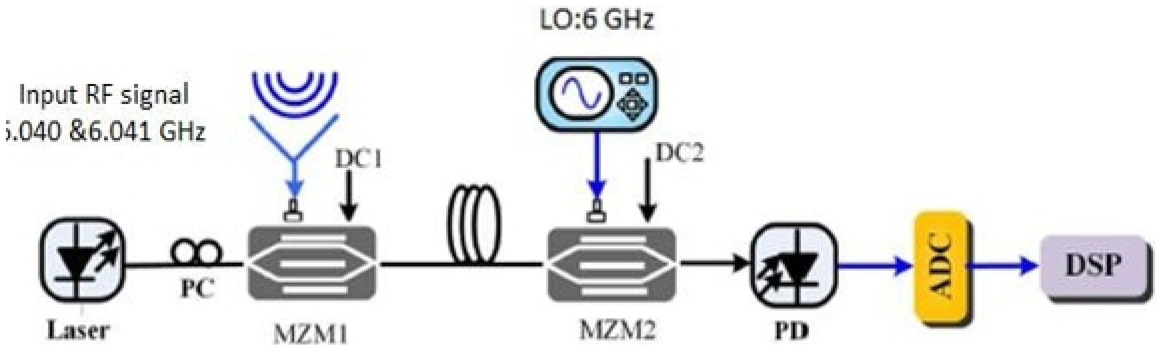

Fig3(a) The schematic diagram[1]

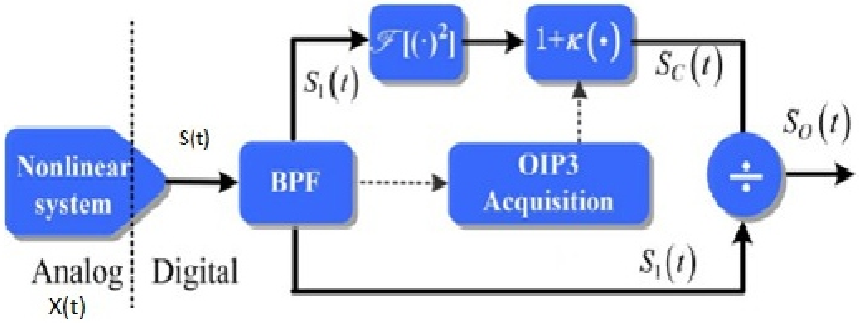

Fig.3(b) The block digram of the proposed Method .

The output of the optical source is modulated by an electrical band-pass signal [1].

$$X\ t\ = A(t)Cos(\omega_{RF}t) \qquad (1)$$

Where $A\ t = 2V(t)\ \cos(\Omega t)$ is the equivalent amplitude of the input two – tone signal.

Here

V (t) is amplitude of input signal.

The RF angle frequency is ω_{RF} .

We apply proposed method .From equation(1) is output of the laser .The output of laser is detected by link. We assume link voltage to voltage transfer function which is amplified by MZM2. We can be expressed in terms of input two tone signal X(t). andgiven by the power series expansion.

$$S\ t\ = C_0 + C_1X + C_2X^2 + C_3X^3 + \ \ldots \ldots \ldots \ldots \ldots \qquad (2)$$

where $C_j\ (j = 1, 2\ , 3\ \ldots .\ )$ are the coefficients of the power series. Which are determined by the specific parameters of the MZM2 and the PD .

Substitute equation (1) into (2), and signal S(t) pass through band pass filter(BPS) , after band pass filter the output voltage around the fundamental signal can be given as

$$S_1\ t\ = C_1\ \ 1 + \frac{3C^3}{4C_1}A^2\ t\ \ .A(t)Cos(\omega_{RF}t) \qquad (3)$$

Band pass filter is pass middle power and stop low power and high power solow order harmonic and higher order harmonic are ignored .

Output of band pass filter S_1 t is through $\mathcal{F}$[(*)2]the narrow-band and low-pass filtering. Than algorithm 1+ƙ(*) .such that as expression is

$$S_c\ t\ =\ 1+\frac{3C^3}{4C_1}A^2\ t\ \ =1+kF\ S_1^2\ t \tag{4}$$

Where ƙ is nonlinear compensation constant.

Here $k=\frac{3C^3}{2C_1^3}$

Result of after algorithm shown in equation (4); its bandwidth should be large enough so that there is no obvious distortion on $A^2\ (t)$ which will be increased when the signal bandwidth is enlarged. However, the bandwidth should be less than $\omega_{RF}\ 2\Pi$. which also shows a limitation on the bandwidth of the input signal. Since the filter is a digital one, its phase response can be ideally linear by a finite impulse response (FIR) filter.

Our proposed method also we note that the third-order output intercept point (OIP3) of the system can be expressed

As

$$OIP3=-\frac{2C_1^3}{3C_3 Z_{out}} \tag{5}$$

where Z_{out} is the output impedance,

Note that the negative sign results fromthe negative *C3*.

We can calculate relation between OIP3 and nonlinear compensation constant. This relation is given as

$$k=-\frac{1}{OIP3\ .Z} \tag{6}$$

The OIP3 should be known by the algorithm, which can be easily and precisely acquired by using a dual-tone signal. The RF spectrum is captured and the corresponding powers of fundamental and IMD3 tones, $P_1$and P_3, can be calculated.it is given as

$$OIP3\ \approx\ \frac{\overline{P_1^3}}{P_3} \tag{7}$$

By using the compensation signal in equation (4), the distortion can be eliminated and the linearized output signal can be finally calculated as

$$S_0\ t\ \approx \frac{S^1(t)}{S_c(t)} = C_1 A(t)\cos(\omega_{IF}\ t) \tag{8}$$

A system containing a pair of modulators, cascaded in series, can be employed to down-convert the RF signals to intermediate frequency .The MZM2 is driven by a strong local oscillator (LO) . The nonlinearity introduced by the MZM2, both difference-frequency and sum-frequency of RF signal and LO are generated. The sum-frequency is beyond the operational bandwidth of the system. Consequently, it doesn't need to be taken into consideration. The presented difference-frequency component is the desired intermediate frequency (IF) signal. After the optical down-conversion, $S_I(t)$is shifted to the IF and additional insertion loss is imposed to it[2].

Our proposed method is made possible by acquiring the OIP3 of the cascaded system therefore The IMD3 induced by the nonlinear voltage to voltage transfer function. We can be eliminated through using digital signal post processing compensation technique, resulting in an improved upper limit for the SFDR..The fundamental to IMD3 ratio is deteriorated once the estimated OIP3 in the digital algorithm deviates from the actual one. The OIP3 will change due to the laser output power fluctuation or link loss variation, which can be updated based on equation (7).

We haveachievedthe best overall NF, a low noise and high gain in this system. The detected IF output is amplified by MZM the photonics link, through which the output level is matched with that of the ADC. So the system is ADC noise dominated while the upper limit of the SFDR can be extended by the proposed method .

This system requires a photodiode capable of handling high optical power. Research devices have been demonstrated that can handle much higher powers than this, but this is still an expensive device. In order to evaluate the performance improvements of the proposed technique. The third-order distortion and noise limited SFDR for this link is derived in dB units per 1 Hz bandwidth by as

$$SFDR = \frac{2}{3}.\ 10\log\left(\frac{2I_{DC}}{e}\right) \tag{9}$$

Where: e is the elementary charge

I_{DC} the effective DC photocurrent.

The Noise figure is calculated by using

$$NF = 10\log\left(\frac{2eV_{\Pi}^2}{I_{DC}\,\Pi^2\,KT\,Z_{in}}\right) \tag{10}$$

Where: V_π is the modulator half-wave voltage,

K is Boltzmann's constant,

T is the system temperature ,

Z_{in} is the input impedance of the system

The Gain is calculated by using

$$G_{dB} = OIP3_{dBm} - \frac{3}{2}SFDR + 174\,\frac{dBm}{Hz} - 10\log B - NF_{dB} \qquad 11$$

Where: B is bandwidth in Hz.

$$OIP3_{dB} = 10\log(OIP3\ W\ \ 0.001\ W).$$

III. Simulation and Results

Result Analysis

We have evaluated Figures of merit and Spurious Free Dynamic Range in Intensity modulation Analog Photonic Link without using any cascade structure of pre and post amplifier. Here we have applied the continuous-wave (CW) laser source which operates at wavelength of 1550 nm is used as the optical carrier. and have generated by two vector signal generators two RF frequencies of 6.040 GHz and 6.041 GHz, and after follow both signal MZM1 . Another polarization controller (PC), which is placed before MZM2, is used to adjust the principal axis of the modulated output of the MZM1 with that of the following MZM2.Now we have applied 6GHz Local Oscillator (LO) and the amplified two-tone RF signal input which is employed to down-convert the RF frequency to IF.

Table 1

Parameter	Value
Input RF signal	6.040 & 6.041 GHz
LO	6 GHz
The effective DC photodiode current (I_{DC})	3 A
The modulator half-wave voltage(V_π)	6 V
The coefficientsC_1 *and*C_3 *respectively*	8 & - 60
The input impedance (Z_{in})	50 Ω
The output impedance(Z_{out})	50 Ω
Thesystem temperature (T)	300 K
Boltzmann's constant (K)	1.381 X $10^{-23}\,J\,K$
Bandwidth in Hz (B)	1 Hz
OIP3	37.5dB
SFDR	130.49 dB in 1-Hz
NF	2.7 dB
Gain	31.08 dB

Simulation

Uur proposed method concept is carried out and shown in Fig. 3. The laser operates at a wavelength of 1550 nm .Two RF tones, with frequencies of 6.040 GHz and 6.041 GHz, respectively, are generated by two vector signal generators and used to drive the modulator. The polarization controller (PC), which is placed before MZM1, is used to adjust the principal axis of the modulated output of the laser with that of the following MZM1. the output modulated light beam is aligned with the following standard MZM2 by another polarized controller. The amplified two-tone input is then introduced to the quadrature biased MZM. Another MZM in series, which is driven by a 6 GHz LO, is employed to down-convert the RF frequency to IF. The modulated optical signal is received by a photo diode (PD) and which makes the output level comparable with that of the following digitizer. The output voltage is then recorded by the ADC (ADC linkwith 14-bit level and 200 MS/s) The digitized signal provided by the ADC then undergoes the digital signal postprocessing(DSP) linearized down converting algorithm as shown in Fig. 3(b) . Firstly, the OIP3 is initialized by using Eq. (6) with a training dual-tone. Consequently, the down-converted signal is obtained and linearized according to Eq. (3) and (7).In order to evaluate the performance improvements of the proposed technique. Thespectram of the proposed technique is compared with that of a conventional link without the proposed technique, which is demonstrated in Fig. 4. The down-converted fundamental at 40.2 MHz and 41.1 MHz as well as the corresponding intermodulation distortions at 39.1 MHz and 42.0Hz are presented for both cases.

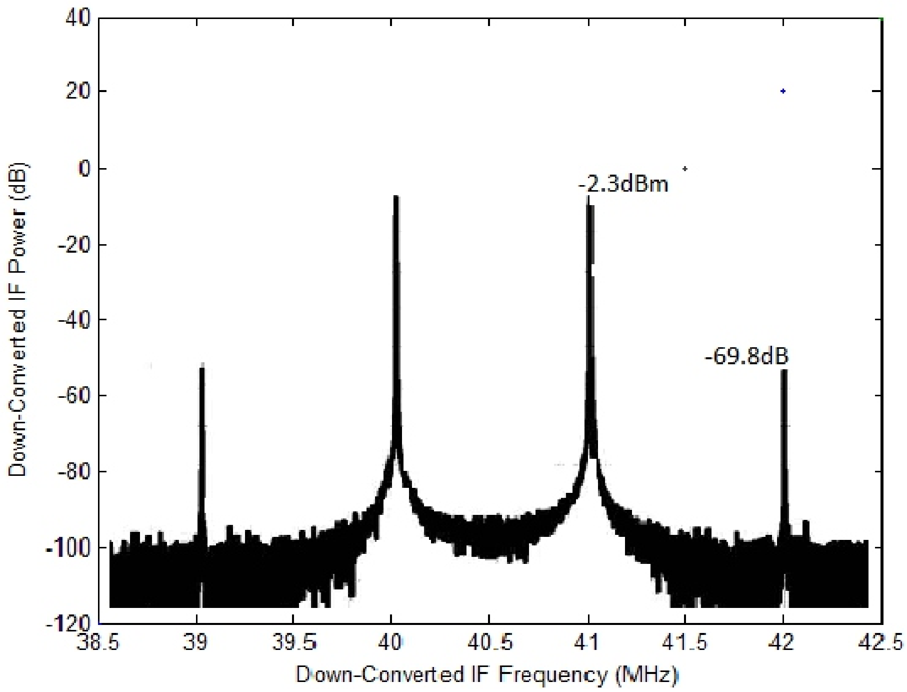

Fig.(a)

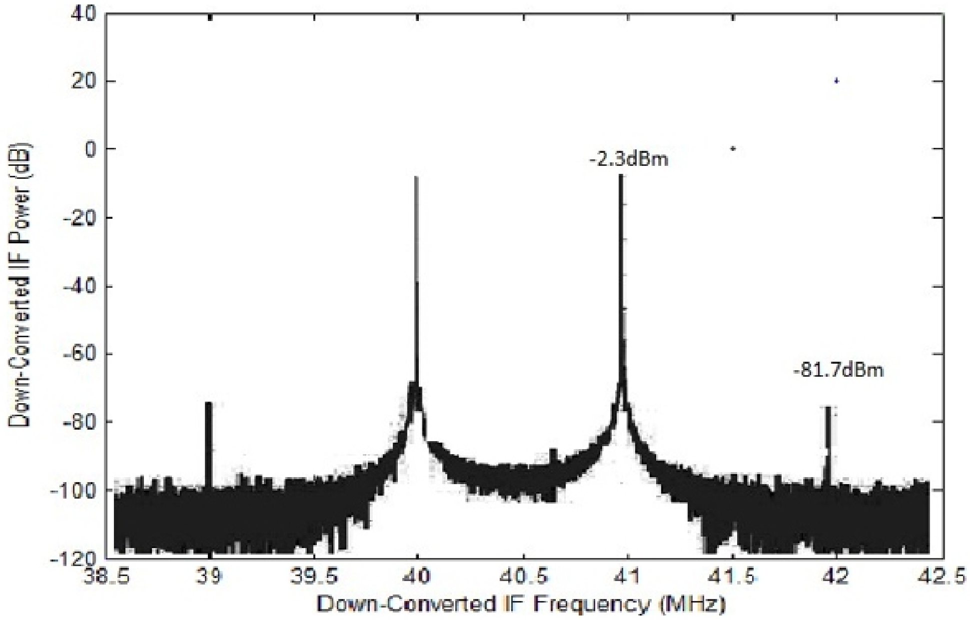

Fig.(b)

Fig. 4. Down-converted electrical spectra (a) before and (b) after the proposed digital signal linearization down conversion algorithm .

The OIP3 used in the algorithm is 37.5dB, which is also the measured value as shown in Fig. 5(a). Fig. 5(a) and 5(b) illustrate the output IF power versus the input RF power at the fundamental frequencies of 6.040 and 6.041 GHz for systems with and without the linearization, respectively.Note that since the proposed approach is performed in digital domain, The IF output is amplified by a specially selected. The down-converted SFDR with nonlinear compensation is 130.49 dB in 1 Hz bandwidth, which is 25.09 dB more than that without compensation, as shown in Fig.5(b).

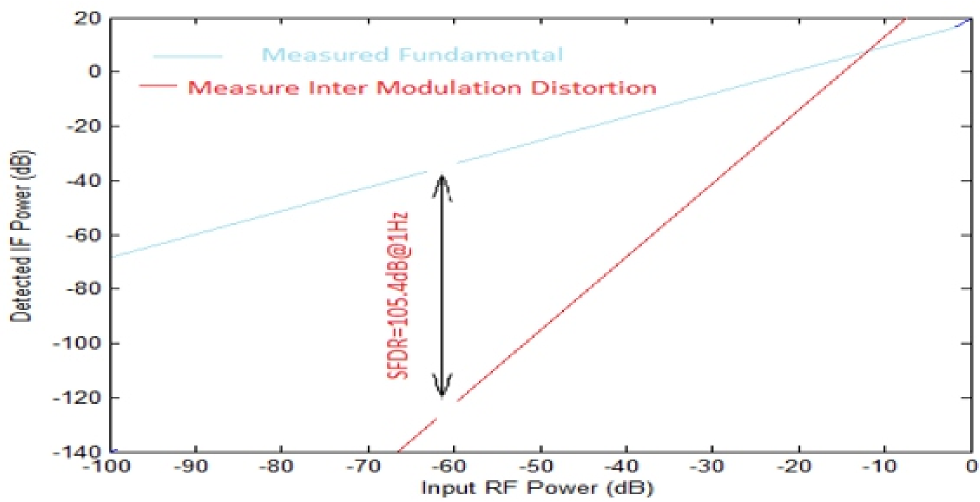

Fig 5.(a)

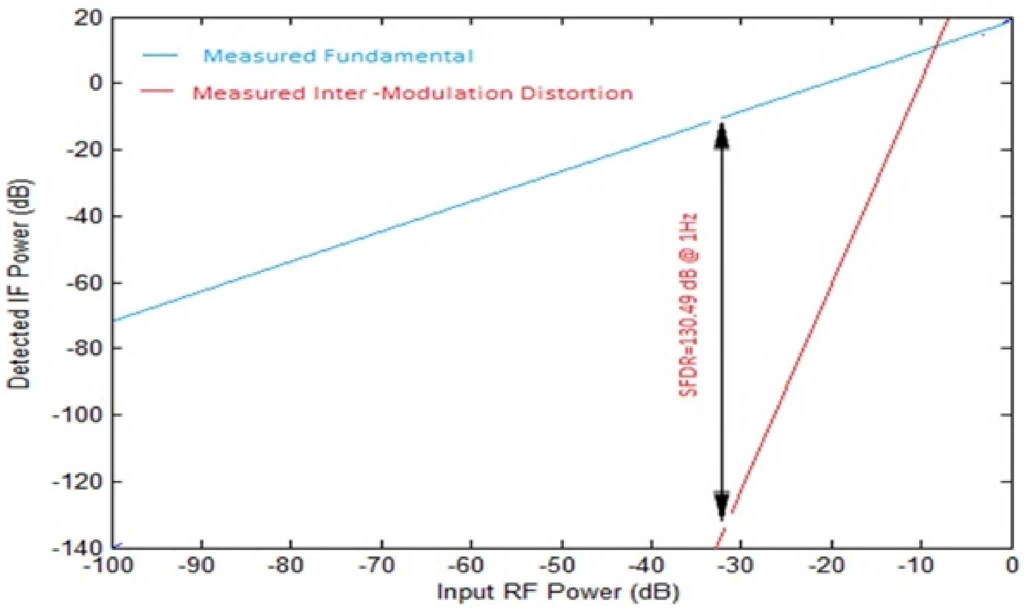

fig.5(b)

Fig. 5 Output fundamental and intermodulation distortion powers versus the input RF power at 6.040 and 6.041 GHz. (a) Before and (b) after the proposed linearized down converting algorithm

IV. Conclusion

We proposed and demonstrated a digital signal for the conventional MZM-based intensity-modulation direct-detection analog photonic link maintaining the noise figure, gain and SFDR same without using any cascade structure of pre and post amplifier. Comparatively measured noise figure and gain of the photonic link are 8.9 and 27.5 dB, respectively and spurious-free dynamic range of 128.3 dB in 1-Hz bandwidth as well as experimental results shows improved noise figure and gain of the photonics link as 2.7 and 31.08 dB, respectively as well as spurious-free dynamic range of 130.49 dB in 1-Hz bandwidth is achieved for a digitizer noise limited analog photonics link without using any cascade structure of pre and post amplifier

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Analytical Study Of Analog Phase-Locked Loop In Message Signals Transmission

Madhavi Marripally[1], Dr Suchi Jain[2]
Department of Electronics and Communication Engineering
[1,2]OPJS University, Churu (Rajasthan)

A B S T R A C T

A phase-locked loop is a criticism framework consolidating a voltage controlled oscillator and a phase comparator so associated that the oscillator frequency (or phase) precisely tracks that of an connected frequency-or phase-tweaked signal. Phase-locked loops can be utilized, for instance, to produce stable yield frequency signals from a settled low-frequency signal. The first phase-locked loops were executed in the mid 1930s by a French architect, de Bellescize. In any case, they just discovered expansive acknowledgment in the commercial center when coordinated PLLs progressed toward becoming accessible as generally minimal effort components in the mid-1960s. The phase locked loop can be dissected as rules as a negative criticism framework with a forward pick up term and an input term.

1. Introduction

A modern propel technology in coordinated circuit technology makes fabrication processes exceptionally appropriate for digital outlines. Little territory and low-voltage plans are ordered by showcase prerequisites. Another favorable position of digital PLL is anything but difficult to update with the procedure changes. Since simple blocks are available in various digital and blended signal ICs, their overhaul is a vital factor in the arrival of another item. Nonetheless, the execution prerequisites of simple blocks requires a total upgrade in another procedure, along these lines expanding the outline process duration.

Decreasing the measure of simple circuitry can enhance the update of these blended signal ICs. A Phase Locked Loop is predominantly utilized with the end goal of synchronization of the frequency and phase of a privately created signal with that of an approaching signal. There are three components in a PLL. The Phase Frequency finder (PFD), the loop channel and the Voltage Controlled Oscillator (VCO). The VCO is the core of any PLL. The component by which this VCO works chooses the kind of the PLL circuit being utilized. There are essentially four sorts of constructing PLLs to be specific Direct PLL (LPLL), Digital PLL (DPLL) and All Digital PLL (ADPLL) [1].

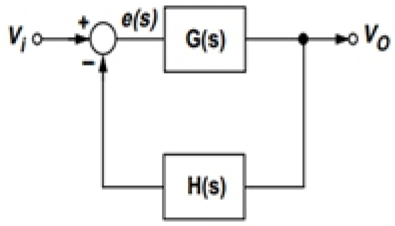

Figure 1. Standard negative-feedback control system model.

In a phase-locked loop, the error signal from the phase comparator is the contrast between the information frequency or phase and that of the signal input. The framework will compel the frequency or phase error signal to zero in the relentless state. The regular conditions for a negative-feedback framework apply.

Forward Gain = G(s), *s = jω = j2πf+

Loop Gain = G(s) × H(s)

Closed Loop Gain G s GsHs − = + ( ) 1 () ()

Because of the integration in the loop, at low frequencies the steady state gain, G(s), is high and VO/VI, Closed-Loop Gain = 1 H

The components of a PLL that contribute to the loop gain include:

1. The phase detector (PD) and charge pump (CP).
2. The loop filter, with a transfer function of Z(s)
3. The voltage-controlled oscillator (VCO), with a sensitivity of KV/s
4. The feedback divider, 1/N

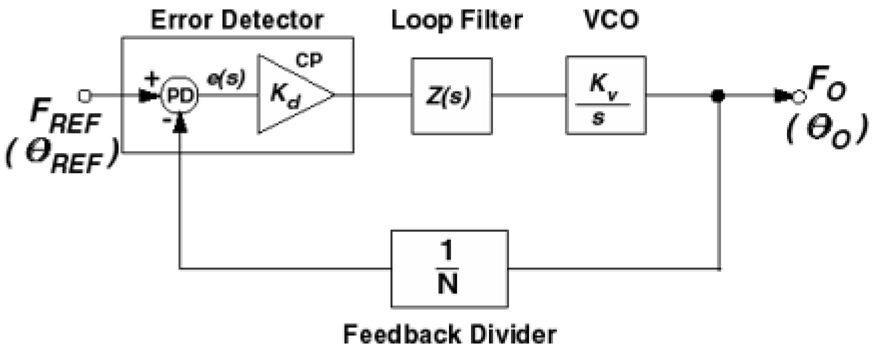

Figure 2.Basic phase-locked-loop model.

In the event that a linear component like a four-quadrant multiplier is utilized as the phase detector, and the loop channel and VCO are additionally simple elements, this is called a simple, or linear PLL (LPLL). In the event that a digital phase detector (EXOR door or J-K flip flounder) is utilized, and everything else remains the same, the framework is known as a digital PLL (DPLL).

In the event that the PLL is fabricated only from digital blocks, with no inactive components or linear elements, it turns into an all-digital PLL (ADPLL) [2].

Finally, with data in digital frame, and the accessibility of adequately quick handling, it is likewise conceivable to create PLLs in the product space. The PLL work is performed by programming and keeps running on a DSP. This is known as a product PLL (SPLL). Alluding to Figure 2, a framework for utilizing a PLL to produce higher frequencies than the info, the VCO sways at an precise frequency of ωD. A bit of this frequency/phase signal is nourished back to the error detector, by means of a frequency divider with a proportion 1/N. This isolated down frequency is nourished to one contribution of the error detector. The other contribution to this illustration is a settled reference frequency/phase. The error detector thinks about the signals at the two sources of info. At the point when the two signal information sources are equivalent in phase and frequency, the error will be zero and the loop is said to be in a "locked" condition. In the event that we essentially take a gander at the error signal, the accompanying equations might be developed [3].

$$e\ s = F_{REF} - \frac{F_0}{N}$$

When

$$e\ s = 0, \frac{F_0}{N} = F_{REF}$$

Thus

$$F_0 = N\ F_{REF}$$

In business PLLs, the phase detector and charge pump together shape the error detector square. At the point when FO ≠ N FREF, the error detector will yield source/sink current heartbeats to the low pass loop channel. This smoothes the present heartbeats into a voltage which thus drives the VCO. The VCO frequency will at that point increment or lessening as fundamental, by KV ΔV, where KV is the VCO affectability in MHz/Volt and ΔV is the change in VCO input voltage. This will proceed until e(s) is zero and the loop is locked. The charge pump and VCO therefore fills in as an integrator, trying to increment or diminishing its yield frequency to the esteem required in order to reestablish its contribution (from the phase detector) to zero.

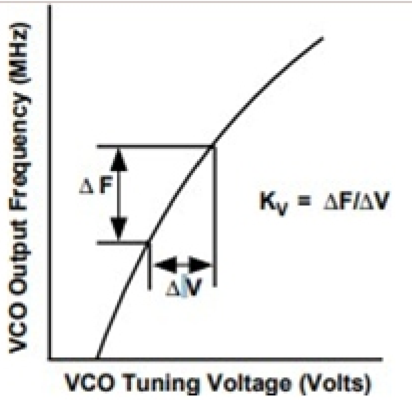

Figure 3. VCO transfer function

The overall transfer function (CLG or Closed-Loop Gain) of the PLL can be expressed simply by using the CLG expression for a negative feedback system as given above.

$$\frac{F_0}{F_{REF}} = \frac{Forward\ Gain}{1 + Loop\ Gain}$$

$$Forward\ Gain,\ G = \frac{K_D K_V Z(s)}{s}$$

$$Loop\ Gain,\ GH = \frac{K_D K_V Z(s)}{N_s}$$

When GH is much greater than 1, we can say that the closed loop transfer function for the PLL system is N and so

$$F_{OUT} = N \times F_{Ref}$$

2. PLL Applications To Frequency Upscaling

The phase-locked loop enables stable high frequencies to be produced from a low-frequency reference. Any framework that requires stable high frequency tuning can profit by the PLL method. Cases of these applications incorporate remote base stations, remote handsets, pagers, CATV frameworks, clock recovery and - era frameworks. A decent case of a PLL application is a GSM handset or base station. Figure 4 demonstrates the get segment of a GSM base station. In the GSM framework, there are 124 channels (8 clients for each channel) of 200-kHz width in the RF band. The aggregate bandwidth involved is 24.8 MHz, which must be examined for action. The handset has a transmit (Tx) scope of 880 MHz to 915 MHz and a get (Rx) scope of 925 MHz to 960 MHz. On the other hand, the base station has a Tx scope of 925 MHz to 960 MHz and a Rx extend of 880 MHz to 915 MHz. For this illustration, we will consider just the base station transmit and get areas. The frequency bands for GSM900 and

Base Station Systems are appeared in Table 1. Table 2 demonstrates the channel numbers for the transporter frequencies (RF channels) inside the frequency bands of Table 1. Fl(n) is the inside frequency of the RF direct in the bring down band (Rx) and Fu(n) is the relating frequency in the upper band (Tx)

Table 1.1: Frequency Bands for GSM900 and Base Station Systems

	T_x	R_x
P-GSM900	935 to 960 MHz	890 to 915 MHz
DCS1800	1805 to 1880 MHz	1710 to 1785 MHz
E-GSM900	925 to 960 MHz	880 to 915 MHz

3. Literature Review

The English physicist Edward Appleton initially portrayed the PLL and it showed up in the Proceedings of the Cambridge Philosophical Society in 1923.In 1953, Gruen distributed a paper particularly on the subject of automatic recurrence control use of PLL in shading TV transmission and gathering. In the year 1979, Gardner, F. examined insights about of PLL. He examined about fundamental standards of phase lock operation, ordinary practices of phase lock building and utilizations of phase lock to different issues. A compact audit of the fundamental PLL standards material to correspondence and control frameworks was displayed by Hsieh, G.C. and, Hung, J.C. in the year 1996.

4. Objectives

The primary target of the proposed work is to display and mimic a moment arrange APLL in time space to ponder the accompanying:

1. To study the transient behavior of phase-locked loop in analog communication
2. To study the Analog Phase Locked Loop in time domain model
3. To study the multiplexing model and its techniques in analog communication
4. To study the dynamic parameters in multiplexing and phase-locked loop in analog communication

5. Research Methodology

The simulations are conveyed outing Turbo C and MATLAB stage. Turbo C is a compiler for C programming dialect from Borland. It was first presented in 1987 and it is noted for its coordinated improvement environment, little size and quick ordering speed. With Turbo C, there is no requirement for a different proofreader, compiler and linker to run the C programs. The primary focal points of C codes are the enhanced precision and adaptability as nonlinear algebraic and differential conditions can be proficiently spoken to. The simulation speed can be essentially expanded by utilizing pre-assembled computer languages, for example, C. As C programming permits abnormal state displaying of useful pieces, along these lines, it is greatly useful for the framework level outline and optimization of framework parameters.

The methodology towards implementation of the proposed work for achieving the objectives as:

1. Derivation of a theoretical model for second order APLL considering different block of elements.
2. Derivation of a theoretical model for circuit level simulation of the LF.
3. Derivation of companion network model for non linear element of the LF based on Backward Euler numeric integration method.
4. Derivation of mathematical model for VCO and multiplier type PD of the model.
5. Derivation of a model for phase error process to study the cycle slips behavior of the model.
6. Derivation of an AWGN model for the APLL.
7. Derivation of a generalized model for the APLL considering different blocks.
8. Simulation of the APLL model by using Turbo C program to study the dynamic characteristics of the PLL and effects of noise on the system.

We are using following tools & methods in our research work:

- Design of an algorithm for the proposed APLL model for simulation in time domain using the Gauss-Seidel iterative method.

The Gauss-Seidel algorithm

1. Take the necessary vectors and arrays as inputs
2. Select an initial guess to start the iteration
3. Solve. Using the Gauss-Seidel Method
4. Check convergence conditions
5. IfConvergence is satisfied go to step 8

- Developing the simulation program incorporating iterative solution options for adjustment of convergence condition, Tsim, T, absolute relative error and tolerance value.

- Developing the simulation program for studying the effects of noise on APLL dynamics.

- Developing the simulation program to study the RMS, histogram and PE) F of phase error behavior.

- Simulating the model to plot the time-domain voltage waveforms of the APLL model.

- Simulating the model to extract the phase and frequency information from time-domain voltage waveforms of the model. The next chapter of the thesis will present in details the theoretical estimation for the proposed APLL model.

6. Results

Phase frequency detector and DCO reproductions are finished utilizing Xilinx and reenactment waveform of same is appeared beneath. The outline has been finished remembering the versatility, adaptability and optimal standard. It can be utilized as a part of any outline suiting the given frequency determinations. A framework clock of 5 MHz is utilized. The plan is executed for a middle frequency of 300 kHz. It's chiefly implied for low frequency applications. The present outline offers a working frequency scope of 290 kHz to 320 kHz around. The outline can be stretched out past that too. Be that as it may, the rationale model should be changed to overcome the proliferation defer that is acquainted due with more noteworthy number of bits engaged with the calculation.

7. Conclusion

As indicated by audit of ADPLL, digital circuits are more adaptable, proficient, adaptable, and less loud as contrasted with simple circuit. This paper talks about the ADPLL configuration utilizing Verilog HDL It likewise displays the FPGA execution in detail. The ADPLL blocks utilized for the outline are additionally given here. This PLL is intended for the middle frequency of 200 kHz and its working frequency scope of ADPLL is 189 kHz to 215 kHz, which is the bolt scope of the plan. At the point when the info was to blame, it consequently continued output arrangement by utilizing the last known right parameters.

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Calculating Total Harmonic Distortion By Measuring Sine Wave

Dr. Paras Nath Singh[1], Er. Prathima Addanki[2], Er. Kumar Anand[3], Er. H. K. Panigrahi[4]

[1]Principal, GIST (Engg. College), Rayagada, Odisha, India
[2]Dept. Of ECE, GITAM University, Vishakhapatnam, Andhra Pradesh, India
[3]Faculty of Engg., Dept. Of EC, MEFGI, Rajkot, Gujrat, India
[4]Dept of Elect. Engg., GIST (Engg. College), Rayagada, Odisha, India

ABSTRACT

Total harmonic distortion is an intricate and habitually confusing concept to grasp. Real-Time direct dig- ital synthesis of analog waveforms using embedded processors and digital signal processors (DSPs) connected to dig- ital-to-analog converters (DACs) is becoming ubiquitous even in the smallest systems. One method to characterize the linearity of an amplifier is to measure its Total Harmonic Distortion (THD). Total Harmonic distortion is meas- ured by utilizing a spectrally pure sine wave in a defined circuit configuration and observing the output spectrum. The amount of distortion present at the output depends on several parameters, such as:

- *Small and large signal nonlinearity of the amplifier being tested*
- *Amplifier"s output amplitude or power level*
- *Frequency response of the amplifier*
- *Load applied to the output of the amplifier*
- *Amplifier"s power supply voltage*
- *Circuit board layout*
- *Grounding*
- *Thermal management, etc.*

Developing waveforms for use in embedded systems or laboratory instruments can be streamlined. One can develop and analyze the waveform generation algorithm and its associated data at desktop before implementing it.

Other important points to consider include algorithm efficiency, data ROM size required, and accuracy/spectral purity of the implementation. Comparing analysis is required when performing own waveform designs. The table data and look-up algorithm alone do not determine performance in the field. Further considerations such as the accuracy and stability of the real-time clock, and digital to analog converter are also required in order to evaluate overall performance. In this paper calculation of THD (Total Harmonic Distortion) is computed & presented by measuring an approximating sine wave.

Keywords: Sine Wave Approximation, Total Harmonic Distortion

1. Introduction

Calculation based on amplitudes (e.g., voltage), measurement must be converted to power to make the addition of harmonic distortion significant. For example, for a voltage signal, the ratio of the square of the RMS voltages is equivalent to the ratio of the power.

Harmonic distortion may be measured by applying a spectrally clean sine wave voltage signal to the input of the amplifier under test[5]. Then, the input power level to the amplifier is adjusted for a de- sired output power level and then looking at the output harmonic spectrums (second, third, and fourth harmonics, etc.) of the amplifier on a spec- trum analyzer relative to the amplitude of the out- put fundamental signal, as shown in Figure 1. Another method is to measure the output wave- form signal (as example shown in Figure 3) of the amplifier using a high speed/bandwidth oscillos- cope (i.e., has BW greater than six to ten times of the fundamental frequency). The power levels of the individual measurement harmonic values (second, third, and fourth, etc.) are usually ex- Total Harmonic Distortion (THD) is expressed in Root-Sum-Square (RSS) in percentage[1,2,3,5]. The THD is usually calculated by taking the root sum of the squares of the first five or six harmon- ics of the fundamental. In many practical situations, there is negligible error when only the second and third harmonics are included, as long as the higher harmonics are three to five times smaller than the largest harmonic. For example, 0.10 is one har- monic value and 0.03 is another higher harmonic value but three times smaller in amplitude:

$$\sqrt{0.10^2 + 0.03^2} = \sqrt{0.0109} = 0.104 \approx 0.1$$

Where P_n|is in watts **(Eq. 1)**

pressed in decibel format, (dBc is relative to the fundamental carrier power level, or dBm is in ab- solute power)[7]. The simplest measurement unit to use for the harmonic measurement is dBm. This allows the tester to not have to keep track of the amplitude signal level of the fundamental fre- quency. For example, if measured in dBc, before calculating the THD, one needs to convert the dBc value to dBm value for each of the harmonic val- ues before calculating their individual power level in watts.

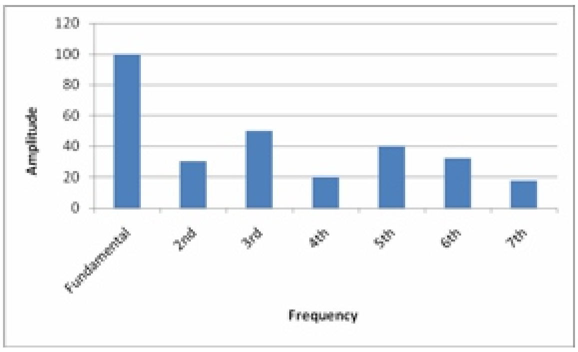

Figure 1: Typical harmonic content of an amplifier's output

or

If the measurement data is in volt,

THD (%) = sqrt($v2^2$ + $v3^2$ + $v4^2$ + …+ vn^2)/v1 *100 where V_n is the RMS voltage (Eq. 2)

Notes:

1. P_n or V_n, where n = harmonic number, n = 1 is the fundamental frequency of the test signal applied to the amplifier under test.

2. Converting the power in dBm to watts:

P(W) = 0.001 ×10P10⁄, (Eq. 3), where P is in dBm.

3. Converting V_{pk} (peak voltage) to V_{RMS} (RMS voltage):

$V_{RMS} = V_{pk}\sqrt{2}$, (Eq. 4)

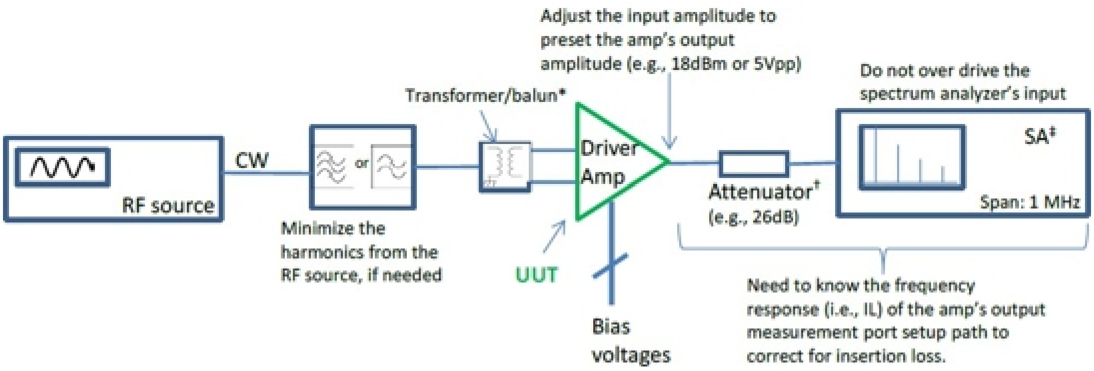

Figure 2: Typical THD Test Measurement Diagram

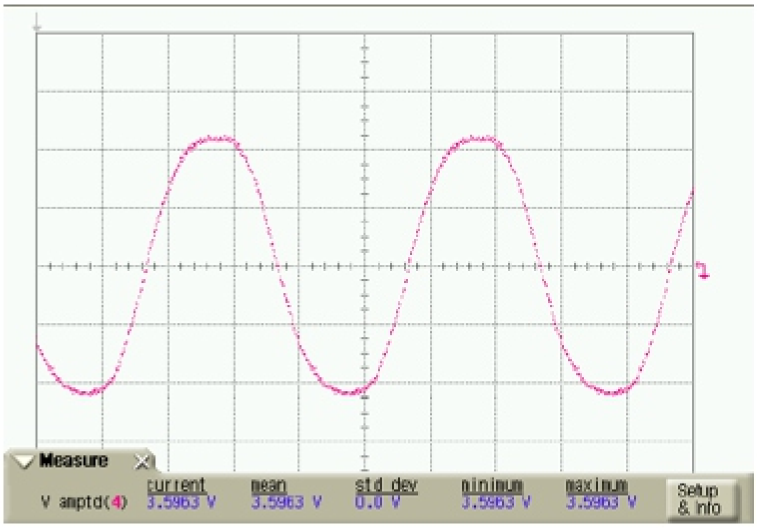

Figure 3: Example of the amplifier‟s output signal at 2.5 GHz, 3.6 Vpp

This demonstration goes through some of the main steps needed to design and evaluate a sine wave data table for use in digital waveform synthesis applications in embedded systems and arbitrary waveform generation instruments.

When feasible, the most accurate way to digitally synthesize a sine wave is to compute the full precision sin() function directly for each time step, folding omega*t into the interval 0 to 2*pi. In real-time systems, the computational burden is typically too large to permit this approach. One popular way around this obstacle is to use a table of values to approximate the behavior of the sin() function, either from 0 to 2*pi, or even half wave or quarter wave data to leverage symmetry.

2. Analysis

A method is used [2,5] a spectrum analyzer to meas- ure the output harmonics of the amplifier (up to the seventh harmonic in the below example), see Figure 2 for the test setup. The following readings were col- lected from the spectrum analyzer, see Table 1:

Freq(GH$_Z$)	Harmonic#	Pout(dBm)	Pout(μW),apply Eq. 3
2.5	1	16.17	41400
5	2	-35.3	0.2951
7.5	3	-6.83	207.5
10	4	-40.6	0.0871
12.5	5	-35.1	0.309
15	6	-37.5	0.1778
17.5	7	-54	0.003981

Table1: Spectrum Analyzer measurement readings

Using equation one above to calculate the THD, (measured at 2.5 GHz and the output voltage is at 3.6 Vpp):

THD (%) = 100 * sqrt((0.2951E-06 + 207.5E-06 + 0.0871E-06 + 0.309E-06 + 0.1778E-06 + 0.003981E-06)/0.0414) = 7.1%

3. Implementation:

A table is created in Double Precision Floating Point. The following commands make a 256 point sine wave and measure its total harmonic distor- tion when sampled first on the points and then by jumping with a delta of 2.5 points per step using linear interpolation. For frequency-based appli- cations, spectral purity can be more important than absolute error in the table. The M-file is used for calculating total harmonic distortion (THD) for digital sine wave generation with or without inter- polation. This THD algorithm proceeds over an integral number of waves to achieve accurate re- sults[4]. The number of wave cycles used is A.

Since the step size 'delta' is A/B and traversing A waves will hit all points in the table at least one time, which is needed to accurately find the aver- age THD across a full cycle.

The relationship used to calculate THD is:
THD = (ET - EF) / ET, % where ET = total energy, and EF = fundamental energy

The energy difference between ET and EF is spu- rious energy. Now the code:

```
N = 256;

s = sin( 2*pi * (0:(N-1))/N)';

thd_ref_1 = ssinthd( s,  1, N, 1, 'direct' )
thd_ref_2p5 = ssinthd( s, 5/2, 2*N, 5, 'linear' )
```

One can put the sine wave designed above into a Simulink model and it can be seen in Figure 4 how it works as a direct lookup and with linear interpo- lation. This model compares the output of the floating point tables to the sin() function. As expected from the THD calculations, the linear interpolation has a lower error than the direct table lookup in comparison to the sin() function.

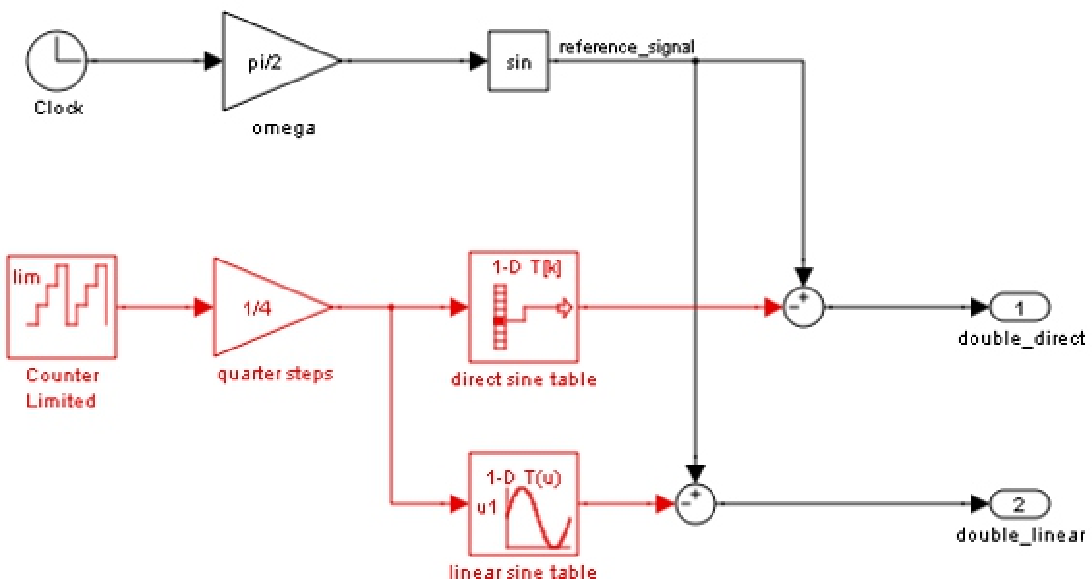

Figure 4: Comparison in approximate sine wave accuracy between direct lookup and linear interpolation vs. reference signal

```
open_system('sldemo_tonegen');
set_param('sldemo_tonegen', 'StopFcn','');
 sim('sldemo_tonegen');
figure('Color',[1,1,1]);
subplot(2,1,1),
plot(tonegenOut.time, tone- genOut.signals(1).values); grid
title('Difference between direct look-up and reference signal');

subplot(2,1,2), plot(tonegenOut.time, tone- genOut.signals(2).values); grid
title('Difference between interpolated look-up and reference signal');
```

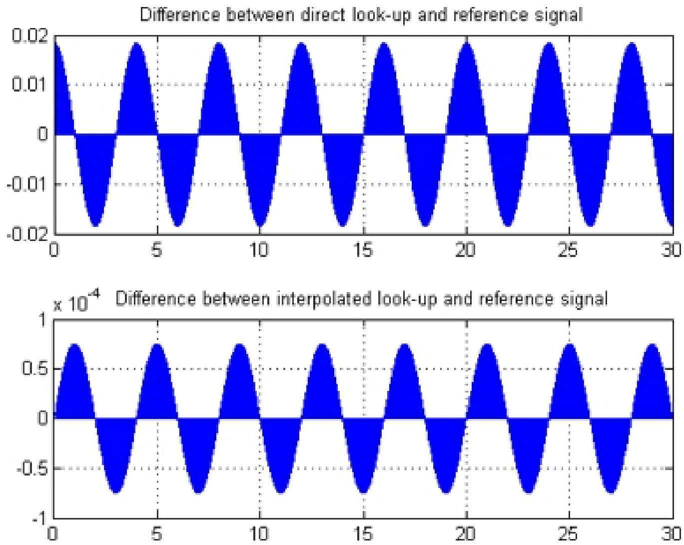

Figure 5: Difference between direct/interpolated look-up & reference signal

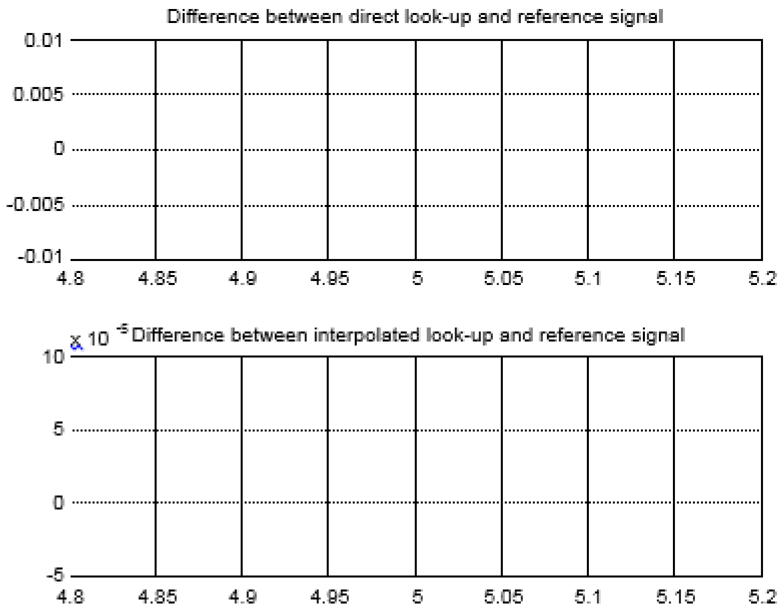

Figure 6: Difference between direct/interpolated look-up & reference signal

Output of the above code is shown in Figure 5 & Figure 6. Taking a Closer Look at Waveform Accuracy and Zooming in on the signals between 4.8 and 5.2 seconds of simulation time (for exam- ple), It can be seen a different characteristic due to the different algorithms used:

```
ax = get(gcf,'Children');
set(ax(2),'xlim',[4.8, 5.2])
set(ax(1),'xlim',[4.8, 5.2])
```

The Same Table is implemented in Fixed Point. Now the floating point table is converted into a 24 bit fractional number using „nearest" rounding.The new table is tested for total harmonic distor- tion in direct lookup mode at 1, 2, and 3 points per step, then with fixed point linear interpolation.

```
Bits = 24;
is  = num2fixpt( s, sfrac(bits), [], 'Nearest', 'on');
thd_direct1 = ssinthd(is, 1, N, 1, 'direct')
thd_direct2 = ssinthd(is, 2, N, 2, 'direct')
thd_direct3 = ssinthd(is, 3, N, 3, 'direct')
thd_linterp_2p5 = ssinthd(is, 5/2, 2*N, 5, 'fixptlinear')
```

Results for different tables and methods are compared. Choosing a table step rate of 8.25 points per step (33/4), it was jumped through the double pre- cision and fixed point tables in both direct and li- near modes and compare distortion results in out- put (Figure 7):

```
thd_double_direct = ssinthd( s, 33/4, 4*N, 33, 'direct')
thd_sfrac24_direct = ssinthd(is, 33/4, 4*N, 33, 'direct')

thd_double_linear = ssinthd( s, 33/4, 4*N, 33, 'linear')
thd_sfrac24_linear = ssinthd(is, 33/4, 4*N, 33, 'fixptlinear')
```

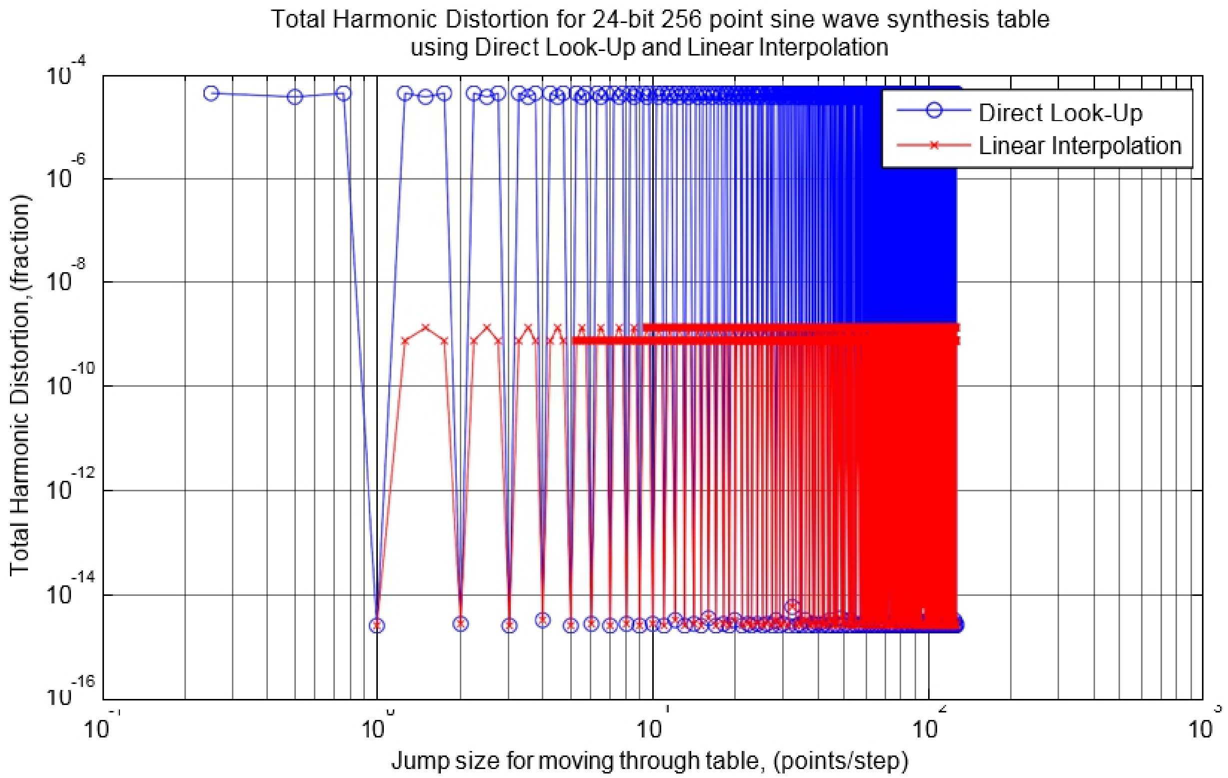

Figure 7: Total Harmonic Distortion

4. Test Cases & Observations:

Using Preconfigured Sine Wave Blocks Simulink also includes a sine wave source block with continuous and discrete modes, plus fixed point sin and cosine function blocks that implement the function approximation with a linearly interpolated lookup table that exploits the quarter wave symmetry of sine and cosine.

Now the model is opened by command:

open_system('sldemo_tonegen_fixpt');

Survey of Behavior for Direct Lookup and Linear Interpolation The script performs a full frequency sweep of the fixed point tables will let us more thoroughly understand the behavior of this design. Total harmonic distortion of the 24-bit fractional fixed point table is measured at each step size, moving through it D points at a time, where D is a number from 1 to N/2, incrementing by 0.25 points. N is 256 points in this example, the 1, 2, 2.5, and 3 cases were done above. Frequency is discrete and therefore a function of the sample rate.

Modes of the distortion behavior in the plot is no- ticed. They match with common sense: when trieving from the table precisely at a point, the er- ror is smallest; linear interpolation has a smallererror than direct lookup in between points. What wasn't apparent from using common sense was that the error is relatively constant for each of the modes up to the Nyquist frequency.

figure('Color',[1,1,1]) tic, sldemo_sweeptable_thd(24, 256), toc

To take this demonstration further, different table precision and element counts are tried to see the effect of each. One can investigate different im- plementation options for waveform synthesis algo- rithms using automatic code generation available from the Real-Time Workshop and production code generation using Real-Time Workshop® Embedded Coder(TM).
bdclose('sldemo_tonegen'); bdclose('sldemo_tonegen_fixpt')

displayEndOfDemoMessage(mfilename) Embedded Target products offer direct connections to a variety of real-time processors and DSPs, in- cluding connection back to the Simulink diagram while the target is running in real-time.

5. Conclusion:

With the use of non-linear loads on the rise global- ly, isolation for poor quality distribution systems and mitigation of harmonics will become increa- singly important. The limits per IEEE Std 519 are not enforced limits but suggestions on acceptable levels. As a result, THD on certain power systems could be much higher, especially considering the difficulty in attaining harmonic measurements [7].

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A Survey On Rainfall Forecasting Using Image Processing Technique

Niyati Salot[1], Dr. Priya R. Swaminarayan[2]

[1]Institute of Computer Application, P.D.Pandya Institute of Computer Application, Gujarat University, Ahmedabad, India.

[2]Professor & Head, MCA Department, Institute of Science and Technology for Advanced Studies & Research Gujarat Technological University, Vallabh Vidyanagar, Gujarat, India.

A B S T R A C T

Rainfall forecasting is important for Agriculture and living things. Agriculture plays an important role in Indian financial system. Agriculture is backbone of our country. Almost 70 percent population is dependent on agriculture. Many industries dependents on agricultural products like sugar industries, Paper industries, tobacco industries etc. Some of the agricultural products like some fruits, vegetables and flowers are exported. Agriculture supports railways and roadways transport. Rainfall is the primary source of water. Human depends on water in daily needs as well as drinking. Water is mainly used in planting trees which are used to bringing natural rainfall. Some agricultural products like rice and corn grown in rainy climate only. Agriculture products are dependents on water so irrigation system is used to provide natural flow of water. Irrigation system is also dependent on rainwater harvesting. Water is essential for the chemical and physiological process in the growth of plant. To find the accurate forecasting, we must know cloud status and sky status. To find the status of cloud, sky and rain required some image processing technique. Image processing is used to enhance images which are taken by Digital camera. It can forecast the behavior of cloud and sky. Image processing has been successfully used by many researchers during last two to four decades. This paper provides a survey of available literature of some methodologies developed by different researchers to forecasting rain by using Image Processing. This survey also reports most suitable technique to forecast rainfall by using Image processing rather traditional statistical and numerical methods.

Keywords : Cloud status Digital Image Processing, Image Processing, Rainfall Forecasting, sky status.

1. Introduction

Accurate rain forecasting is the most important for an agricultural and water resource. All living things need water to leave. It is very important to maintain water-cycle and keep proper ecological condition. In the, normal water-cycle rain water goes into the earth. In the earth water moves and reaches it into lakes, sea or ponds. And that water is again lifted out by sun using high elevation as cloud. When cloud reaches at certain height it will cooled and converted into water vapor. Due to gravity it again comes on to the earth. By forecasting rainfall cropping system can be planned according to rainfall pattern. Rainfall forecasting is useful in taking decisions on time of sowing, planning of irrigation, time of harvesting. It is also useful to designing farm ponds, tanks or irrigation projects. In crop production system important aspects of rainfall are its amount, distribution and intensity.

Many different researchers developed different model of rain forecasting by using newer technologies. Rainfall is of non-liner nature so that accuracy of forecasting is in satisfactory level. Image processing technique is an attractive approach to forecast rainfall because of it keeps consistently high quality of Image. It is also low cost processing. It has ability to manipulate all aspects of the process. In this processing images are effectively stored and efficiently transmitted. Image processing has been successfully used in this numerous approaches like Image denoisy, Watermarking, Medical Imaging, and Brain-computing.

II. Concept of Image Processing

Image processing is a technique to enhance raw images taken by digital camera, satellite, space probes or air craft for various applications. Image processing systems are developed due to easy powerful enhance personal computer available, more storage device and graphic software available. Image processing is used in many applications such as:

- Medical Imaging
- Remote Sensing
- UV Imaging
- Robot Vision
- Hurdle Detection
- Pattern Recognition
- Video Processing

The Image processing common steps are: Image Scanning, Image storing, enhancing and interpretation.
[1]

II. Methodology

The various Image processing methods are: Image storage and manipulation, enhancement, restoration, analysis, reconstruction, compression.

3.1. Methods to be used in processing data

3.1.1. Image storage and manipulation

In this method, First of all Images are stored in to different format. In earlier window operating system, Images are stored in bitmap format. In UNIX operating system, Images are stored in raster file format. After saving the images different functions are applied to read the images.

3.1.2. Image enhancement

In this method different methods are applied on to images to enhance its visual appearance which is easier to machine or human interpretation. Image enhancement examples are noise filtering, contrast and edge enhancement, pseudo-coloring, sharpening, and magnifying. In Image enhancement process only some characteristics added. These algorithms are generally interactive and application specific. Some enhancement techniques are: contrast stretching, Noise filtering and Histogram Modification. [2]

3.1.3. Image restoration

In this method, degradation or noise is removed. In this method some functions are applied like de-blurring, noise filtering and correction of distortion.

3.1.4. Image analysis

In Image analysis quantitative measurements are derived in to description [3]. Image Analysis technique includes some extraction feature that will used to identify object. Some of its usages are quantitative information used to take sophisticated decision like robot moves his head. Quantitative measurements of objects allow classifying of images.

3.1.5. Image reconstruction

In Image reconstruction method two – dimensional images are converted into single dimension projection. Each projection is calculated by passing X-ray beams from an object. Reconstruction algorithm derives a thin axial slice of object; allow viewing inside the object without cutting. Such techniques are used in medical Imaging.

3.1.6. Image compression

Image compression coverts image into some data which can be transmitted on to the network. There are two mainly techniques for image compression: Lossy compression and lossless compression. In recent year wavelet based compression techniques are used for higher compression ratio.

3.2. Parameters

Most of the studies in rainfall forecasting are primarily based on statistical and empirical techniques. For the forecasting rainfall studies of historical data sets have brought out several predictors. These parameters represent different components like sky, cloud and its shape, color and type. In this section, the characteristic features of the known predictors grouped into appropriate categories are described.

3.2.1. Status of Sky

To find status of sky we will be used either texture description method or wavelet. In that method we can identify clouds and distance of them.

3.2.2. Status of cloud

The clouds are changed according to its thickness. High density will be detected. Middle level density will be known as sky. Mild time density will be known as cloud. The highest weight will be considered as sky and lower weight will be considered as cloud.

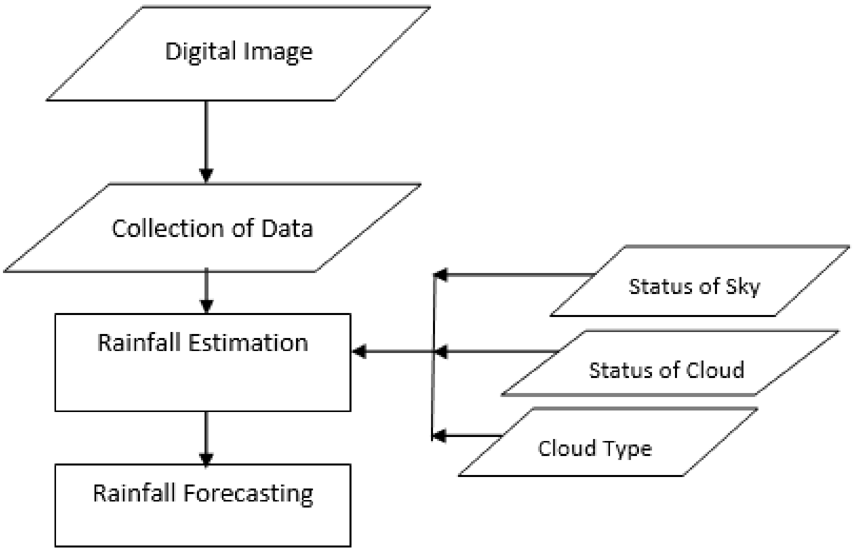
3.2.3. Types of cloud

To find the types of cloud, cloud status is used because cloud has different shapes and different density. These values will be matched accordingly in different algorithm. Different concepts like color, texture and shape used to find the types of cloud.

3.3. Proposed system Design

The proposed system design steps are as follows:

- In the first step, the clouds images are taken by digital camera are processed by Image processing software.
- From the digital images information is collected in binary format. So that Images data can be read and write.
- To forecasting rainfall three parameters are used with the Image data. Three parameters are: Status of sky, Status of cloud and cloud type.
- After combining all the data are inputted to algorithm and that will find accurate rainfall estimation.



IV. Literature review

YanlingHao, Wei ShangGuan, Yi Zhu, and YanHong Tang [4] describes the method Content- based cloud image processing and information retrieval(CBIPR) for finding texture for clouds. In this method satellite Images of the clouds are used. It has the features of texture of clouds. This method describes basic character, like cloud's color, texture, edge and shape from the cloud image. In this method satellite images of the clouds are stored into the database and database will provide basic characteristics.There are some limitations in traditional image capturing method for realized image retrieval accurately and quickly. So, CBIPR method is adaptive. This method is very much useful for finding cloud texture, size and color. It gives accurate result.

Peter S. Masika[5] describes an algorithm for Cloud height determination and comparison with observed rainfall by using meteosat second generation (msg) imageries. In this method we can obtain approximations of surface and cloud parameters from satellite data. This algorithm developed to identify cloud-free and cloud-contaminated pixels. It will find accurate shape of clouds. This study attempts to compare computed cloud height and observed rainfall on ground station. It will derive relationship between cloud height and total rainfall from storms on the same station. The study starts from the foundation of a suitable balance between calculation speed and exactness in the output data. By using cloud masking accuracy found 87%. They give accurate result from cloud height and total rainfall. Based on the developed simple cloud mask algorithm; it was found that setting thresholds for transmission all cloudy pixels in satellite images is the most difficult part in threshold techniques. The main problem is that the thresholds are functions of many parameters such as surface type (land, ocean, ice), surface conditions (vegetation, soil vapor), current weather (which changes surface temperature and reflectance significantly), atmospheric conditions (temperature inversions, haze, foggy), season, time of day and even satellite-earth-sun geometry (hence bidirectional reflectance and sun glint).In this paper they developed simple cloud mask algorithm but it has some limitations like there were always need for spatial and temporal averaging data. In order to gain accurate result further threshold tests are required based on cloud microphysical process on cloud particles.

Shou Yixuan, Li Shenshen, Shou Shaowen and Zhao Zhongming [6] describe Infrared images of the clouds give information of the texture of clouds and also give the rainfall estimation. In that article they have describe the application of a classification method significant four features. Four features are: energy, entropy, inertial-quadrature and local calm. They have use neural network classification method for textural features. In addition, method identifies relationship between textural classification and rainfall. So this scheme is helpful for weather analysis and rain forecasting. This paper provides methods which is enough to whether analyzing and forecasting. But it has some limitations that it required correlative methods to compare the data such as numerical method.

Wei Shangguan; YanlingHao; Zhizhong Lu; Peng Wu [7] states that the research development of satellite cloud image processing technology become very quickly and it also concentrate on judge the cloud type and classification of cloud. In this study, different techniques are used such as image processing and pattern recognition. In satellite cloud image, cloud structure is very important feature. Since, satellite image has clear cloud structure and computer texture analysis provide perfect future for study and analyze all kinds of satellite images. In this paper, computer image texture analysis technology combines with variation theory. It will extract and analyze the texture feature of similar cloud and clear sky.

K. Richards and G.D. Sullivan [8] describes the methods for using color and texture to discriminate cloud and sky in images captured using a ground based color camera. But in this approach method alone does not proved sufficient differentiate between different types of cloud and sky. This classification can develop by combining the features using a Bayesian scheme.

Malay K. Kundu and PriyankBagrecha [9] describe the M-band Wavelet Transform based feature extraction algorithm is explained in his paper. The MxM sub-bands are used as basic features. In which energies computed in a neighborhood are taken as the features for each pixel of the image. These features are clustered using FCM to get image signature for similarity matching using the Earth Mover's Distance. The results acquired were matched with MPEG-7 content descriptor based system and found to be superior.

Wei Shangguan; YanlingHao; Kuo-Lin Hssu1, X. Gaoand Soroosh Sorooshian1 [10] describes the method for identifying cloud texture classification mapping using satellite images. This approach is used to identifying height of cloud and cloud texture. In this scheme firstly classifying cloud types based on texture features of regional cloud images. The output will regressing the relationship of cloud brightness temperature and surface rain rate depend upon different cloud types. In this method digital images are used so it can be taken up by digital camera so it is less costly. But this method has some limitation like it does not give accurate result. For getting accurate result variation technique must be used.

V. Conclustion

This paper reports a detailed survey on rainfall forecasting using Image processing techniques. Some of the researchers used satellite images of cloud and sky. And some of the researcher used digital image processing technique which is cheaper technique. Researchers have applied different algorithm on cloud and sky images. However in these methods some of the drawbacks have been founded. This paper is used in reference for the further development of rainfall forecasting by using image processing technique.

The survey also gives a conclusion that the forecasting techniques that use Content- based cloud image processing and information retrieval (CBIPR), meteosat second generation (msg) imageries, M-band Wavelet Transform based feature extraction algorithm. By this literature review some researcher developed adaptive method which will be helpful to me for my research.

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A Comprehensive Study Of Facts About Challenges And Future Direction For MIMO In Communication System

Krishna Murthy Ajmera[1], Dr. Phool Singh Chouhan[2], Dr. Senthilkumar A[3]
Department of Electronics andCommunication Engineering
[1,2,3]OPJS University, Churu (Rajasthan)

ABSTRACT

In this paper primary goal is to provide the reasonable issues experienced in MIMO communication system. Here just two noteworthy issues are considered. The first is the estimation of the channel state data and the impacts of the estimation blunder on the system execution. The second subject point by point is the conceivable sub-channel relationship between's transmit and get antenna sets and the execution examination with the nearness of connection. One of the principle contemplations in MIMO communication system is antenna choice which is likewise examined in this paper. Couple of new technology Multi-Input Multi-Output (MIMO) is the Future Wireless system will be a great deal more productive to take care of the substantial demand of Wireless communication in accessible constrained recurrence resources.

Keywords: 4G; LTE; MIMO; MISO; SIMO; SISO

1. Introduction

Wireless system keep on striving for ever higher information rates. This objective is especially trying for frameworks that are power, transmission capacity, and intricacy restricted [1]. Be that as it may, another area can be abused to fundamentally expand channel limit: the utilization of various transmit and get receiving wires. This report condenses the fragment of the late work concentrated on the limit of MIMO frameworks for both single clients and numerous clients under various presumptions about the spatial relationship and channel data accessible at the transmitter and collector.

The huge phantom efficiencies connected with MIMO channels depend on the preface that a rich dissipating environment gives autonomous transmission ways from each transmit receiving wire to each get reception apparatus. In this manner, for single-client frameworks [2], a transmission and gathering system that adventures this structure accomplishes limit on around min(M,N) isolate channels, where is the quantity of transmit reception apparatuses and N is the quantity of get radio wires. In this manner, limit scales directly with min (M,N) in respect to a framework with only one transmit one get radio wire.

his limit increment requires a disseminating situation to such an extent that the framework of channel picks up amongst transmits and gets reception apparatus sets has full rank and free passages and that flawless appraisal of these additions are accessible at the beneficiary [3]. Idealize appraisals of these additions at both the transmitter and recipient give an expansion in the steady multiplier connected with the straight scaling. MIMO channel limit depends vigorously on the factual properties and receiving wire component relationships of the channel.

2. Mathematical Model Of MIMO

Consider a remote correspondence framework with Nt transmit (TX) and Nr gets (RX) receiving wires. The thought is to transmit diverse floods of information on the distinctive transmit radio wires, yet at similar bearer recurrence [5]. The stream on the p-th transmits receiving wire, as the capacity of the time t, will be indicated by sp (t). At the point when a transmission happens, the transmitted flag from the p-th TX receiving wire may discover diverse ways to land at the q- th RX reception apparatus, in particular, an immediate way and circuitous ways through various reflections.

$$( ) = \Sigma \quad h \; ( ) \; ( ),$$

On the off chance that quantity of conditions is bigger than quantity of questions, an answer can be found by playing out a projection utilizing the slightest squares technique [4], otherwise called the Zero Forcing (ZF) strategy. For the symmetric case, the ZF arrangement brings about the novel arrangement.

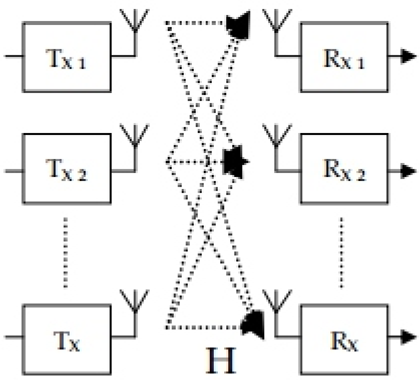

Figure 1: Schematics Representation Of A MIMO Communication System

3. Generation And Communication Models

Here we will study different generation from first upto LTE:

1G: 1G (or 1-G) alludes to the original of wireless stele phone technology, mobile communications.

These are the simple telecommunications models that were presented in the 1980s and preceded until being supplanted by 2Gdigital telecommunications. 1G speed change between that of a 28k modem (28kbit/s) and 56k modem (56kbit/s), which means genuine download paces of 28kbit/s to 56kbit/s Antecedent to 1G technology is the mobile radio phone, or 0G.

2G: 2G (or 2-G) is short for second-generation wireless stele phone technology. Secondgeneration 2G cell telecom networks were industrially propelled on the GSM standard in Finland by Radiolinja (now a portion of Elisa Oyj) in 1991 [5]. Three essential advantages of 2G networks over their antecedents were that telephone discussions were carefully scrambled; 2G systems were altogether more proficient on the range taking into account far more prominent mobile telephone infiltration levels; and 2G presented information administrations for mobile, beginning with SMS instant messages. 2G network takes into account much more noteworthy entrance power. 2G innovations empowered the different mobile telephone networks to give the administrations, for example, instant messages, picture messages and MMS (multi media messages). All instant messages sent more than 2G are carefully scrambled, taking into account the move of information in a manner that exclusive the proposed receiver can get and read it.

3G: 3G, short for third Generation, is the third generation of mobile telecommunications technology. 3G telecommunication networks bolster benefits that give a data exchange rate of no less than 200 kbit/s. Later 3G discharges, regularly meant 3.5G and 3.75G, likewise give mobile broadband access of a few Mbit/s to cell phones and mobile modems in PCs.

4G: 4G, short fourth generation, is the fourth generation of mobile telecommunications technology succeeding 3G. A 4G system, notwithstanding regular voice and different administrations of 3G system, gives mobile ultra- broadband Internet access, for instance to portable workstations with USB wireless modems, to advanced cells, and to other mobile gadgets.

5G: 5G (fifth generation mobile networks or fifth generation wireless systems) indicates the following real period of mobile telecommunications models past the current 4G/IMT-Advanced principles. 5G is additionally alluded to as past 2020 mobile communications advancements. 5G does not depict a specific detail in any official report distributed by any telecommunication standardization body. In spite of the fact that upgraded benchmarks that characterize capacities past those characterized in the current 4G guidelines are under thought, those new abilities are as yet being assembled under the current ITU-T 4G gauges.

LTE: LTE, an acronym for Long Term Evolution, regularly advertised as 4G LTE, is a standard for wireless communication of rapid information for mobile telephones and information terminals. It depends on the GSM/EDGE and UMTS/HSPA network advancements, expanding the limit and speed utilizing an alternate radio interface together with center network upgrades.

Presently, basically there are four communication models in wireless communication which are depicted as takes after:

SISO: SISO remains for single input single output. In this communication single antenna in accessible at transmitter and single at receiver, SISO systems are normally less mind boggling than

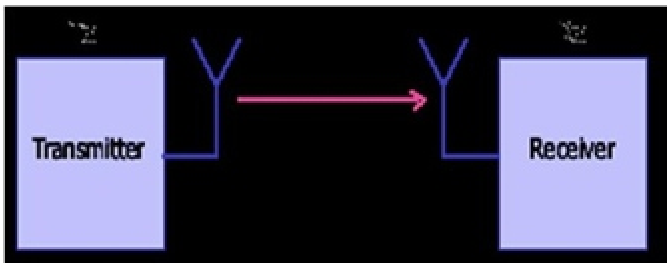

Figure 2 SISO Model

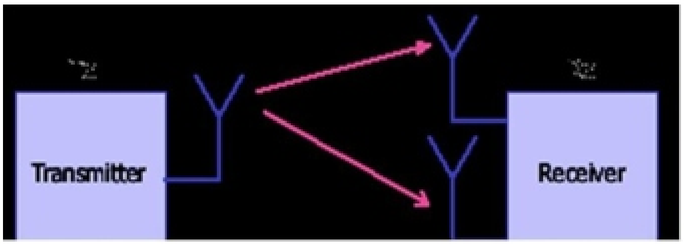

Figure 3 Single Input Multi Output

Multiple-output (MIMO) systems. More often than not, it is likewise less demanding to make request of greatness or inclining forecasts "on the fly" or "back of the envelope". MIMO systems have an excessive number of connections for a large portion of us to follow through them rapidly, altogether, and viably in our heads.

SIMO: SIMO remains for single input multiple outputs. In this communication demonstrate single antenna is associated at transmitter and multiple at accepting side. The wireless communications in which multiple antennas are utilized at the goal (receiver). The antennas are joined to minimize blunders and upgrade information speed.

MISO: MISO remains for multiple input single output. In this model multiple antennas are associated at transmitter and single at the receiver. MISO (multiple inputs, single output) is an antenna technology for wireless communications in which multiple antennas are utilized at the source (transmitter). The antennas are consolidated to minimize mistakes and enhance information speed. The goal (receiver) has just a single antenna. MISO is one of a few types of shrewd antenna technology, the others being MIMO (multiple inputs, multiple outputs) and SIMO (single input, multiple output).

MIMO: MIMO remains for multiple input multiple output. In this communication show

4. History And Principle Of MIMO

multiple antennas are at transmitter and multiple at receiver side because of which they got signal get change and this further enhances the throughput of communication model.

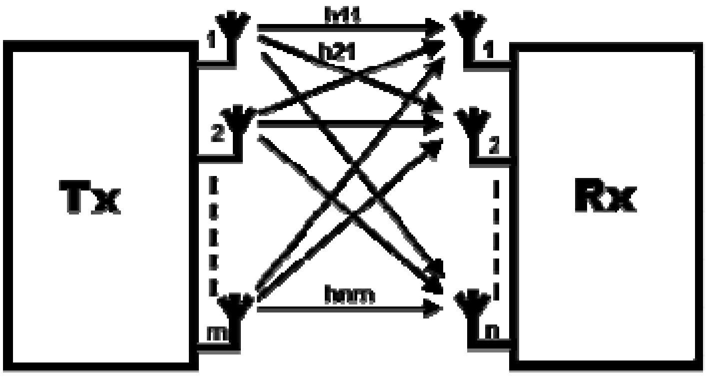

Figure 4 Multi Input Single Outputs (MISO)

History

The most punctual thoughts in this field backtrack to work by AR Kaye and DA George (1970), Brandenburg and Wyner (1974) and W. van Etten (1975, 1976). Jack Winters and Jack Salz at Bell Laboratories distributed a few papers on bar framing related applications in 1984 and 1986.

Principle

The fundamental standard behind this technology MIMO utilizes technology Named as spatial multiplexing and bar framing. In 1996, Greg Raleigh, Gerard J. Foschini, and Emre Telatar refined new ways to deal with MIMO technology, considering an arrangement where multiple transmit antennas are co-situated at one transmitter to enhance the link throughput successfully [6–8]. Chime Labs was the first to exhibit a research center prototype of spatial multiplexing in 1998, where spatial multiplexing is a key technology to enhance the execution of MIMO communication systems [9].

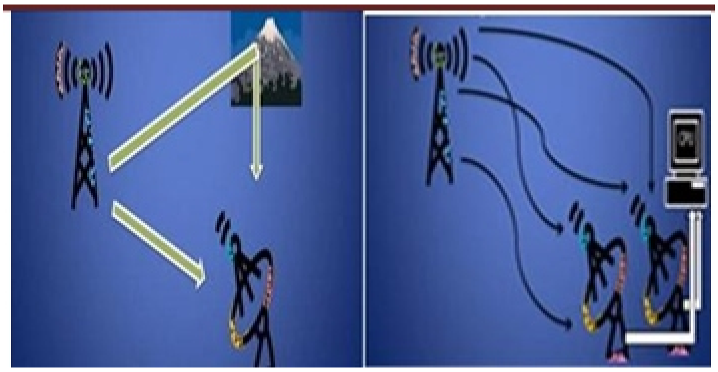

Figure 5. Spatial multiplexing with and without

5. MIMO System

MIMO is fundamentally a technology in which multiple antennas are utilized both at transmitter and receiver to enhance communication execution. It is one of a few types of shrewd antenna technology. Here the term input and output alludes to the radio channel conveying the signal not to the gadget having antenna. Fundamental goal is to contemplate the essential guideline behind this technology. MIMO utilizes technology named as spatial multiplexing and pillar framing.

Spatial Multiplexing: - In spatial multiplexing, a high rate signal is part into multiple lower rate streams and every stream is transmitted from an alternate transmit antenna in a similar recurrence channel. In this signal originating from various antenna are joined and suitable numerical capacity is connected to get adequate output. Underneath two fig demonstrates spatial multiplexing with and without SM which essentially lets us know that in first visual the output got at getting antenna is with and without defers that is unquestionably not the proficient output to endless supply of one in second visual which delineates that output acquired at PC are certainly a great deal more grounded and further can be enhanced by applying suitable scientific capacity as per the kind of use.

Shaft shaping: - Beam framing or spatial sifting is a signal handling system utilized as a part of sensor clusters for directional signal transmission or gathering. This is accomplished by consolidating components in a staged exhibit in a manner that signals at specific edges encounter valuable obstruction while others encounter ruinous impedance. Shaft shaping can be utilized at both the transmitting and getting closes keeping in mind the end goal to accomplish spatial selectivity. The change contrasted and omnidirectional recep-tion/transmission is known as the get/transmit pick up (or misfortune).

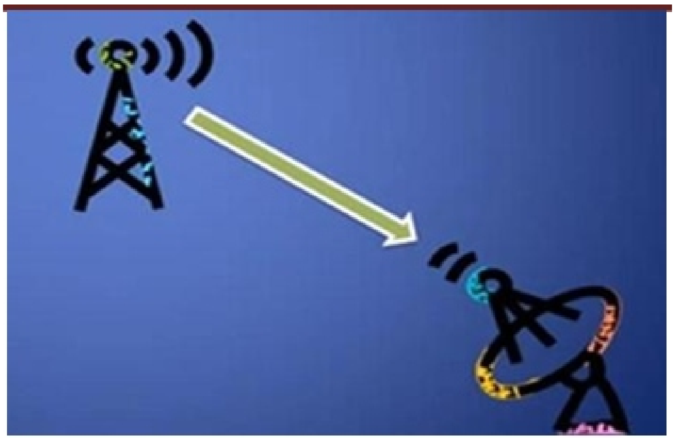

Figure 6. Beam forming

Gadget call cooperatively transmit the same subcarrier which ca likewise twofold uplink limit. To enhance the information rate or throughput of wireless get to even under state of impedance, signal blurring for long separation along this utilization of restricted bandwidth successfully. In the event that the impediments are to be considered of MIMO system are fundamentally its planning, multichannel synchronization, DSP specialists are required to execute more refined baseband handling calculation to better decipher the channel demonstrate.

6. Capacity of MIMO

MIMO can be sub-parceled into three central classes, precoding, spatial multiplexing or SM, and assorted qualities coding.

Spatial multiplexing requires MIMO radio wire setup. In spatial multiplexing, a high rate flag is part into different lower rate streams and each stream is transmitted from another transmit radio wire in a comparable repeat channel. If these signs connect at the recipient receiving wire display with satisfactorily special spatial imprints and the beneficiary has exact CSI, it can disconnect these streams into (for all intents and purposes) parallel channels. Spatial multiplexing is a viable strategy for growing channel confine at higher Signal-to-commotion extents (SNR). The most outrageous number of spatial streams is confined by the lesser of the amount of receiving wires at the transmitter or recipient. Spatial multiplexing can be used without CSI at the transmitter, yet can be joined with precoding if CSI is available. Spatial multiplexing can moreover be used for synchronous transmission to numerous collectors, known as space-division various get to or multi-customer MIMO, in which case CSI is required at the transmitter. The booking of collectors with different spatial imprints grants incredible distinctness.

Differences coding techniques are used when there is no channel learning at the transmitter. In differing qualities techniques, a singular stream (unlike various streams in spatial multiplexing) is transmitted, yet the flag is coded using strategies called space-time coding. The flag is transmitted from each of the transmit reception apparatuses with full or close orthogonal coding Diversity coding manhandle the free obscuring in the numerous radio wire connections to update flag assorted qualities. Since there is no channel learning, there is no pillar shaping or show get from assorted qualities coding. Assorted qualities coding can be combined with spatial multiplexing when some channel data is available at the transmitter.

Moreover, speculative examination exhibited that reception apparatus assurance keeps up the high data rate of spatial multiplexing MIMO frameworks, and improves assorted qualities orchestrate in every information stream without complex space-time taking care of at transmitters and recipients [13]. Like most designing issues, the arrangement of receiving wire decision count incorporates numerous, regularly clashing layout criteria and finding the perfect arrangement is thusly not a direct errand. Subsequently, there is a necessity for headway based system that can be used to layout receiving wire subset decision techniques for MIMO-OFDM. In this way, finding a genuine perfect radio wire choice plot which satisfies concentrated on QOS, improved point of confinement and transmission unwavering quality prerequisites is still a test.

As of late, authorities have proposed a couple of systems for perfect radio wire subset choice in MIMO-OFDM and can be broadly requested into three classes.

7. Antenna Seelction For Mimo

The huge impediment in MIMO-based frameworks is the cost of the gear, in light of the fact that each a tennaelement requires an aggregate radio repeat (RF) chain that is made out of blenders, intensifiers, and easy to- automated change. A promising technique named reception apparatus subset decision an alluring plan has been proposed to decrease the gear unusualness, i.e., spare cash on RF chains, while holding various arranged points of interest [10, 11].

In addition, it staggeringly improves the throughput/enduring quality tradeoff [12]. In such subset decisions, the amount of RF chains is more diminutive than the genuine number of radio wire segments. The RF chains are connected with the "best" receiving wire parts.

What's more, theoretical examination showed that radio wire assurance keeps up the high data rate of spatial multiplexing MIMO frameworks, and upgrades differing qualities orchestrate in every information stream without complex space-time taking care of at transmitters and collectors [13]. Like most building issues, the arrangement of radio wire decision count incorporates various, regularly clashing layout criteria and finding the perfect arrangement is thusly not a direct assignment. Subsequently, there is a prerequisite for headway based technique that can be used to framework receiving wire subset decision methodology for MIMO-OFDM. In this way, finding a genuine perfect reception apparatus choice plot which satisfies concentrated on QOS, improved farthest point and transmission dependability necessities is still a test.

As of late, pros have proposed a couple of strategies for perfect receiving wire subset determination in MIMO-OFDM and can be widely requested into three classes.

For example,
1. Transmit Antenna Selection (TAS)
2. Receive Antenna Selection (RAS)
3. Joint Antenna Selection (JAS)

8. Chalanges Faced By MIMO

Channel State Information Estimation: In wireless communications, channel state data (CSI) alludes to known channel properties of a communication link. This data portrays how a signal spreads from the transmitter to the receiver and speaks to the joined impact of, for instance, dispersing, blurring, and control rot with separation. The CSI makes it conceivable to adjust transmissions to current channel conditions, which is pivotal for accomplishing dependable communication with high information rates in multi-antenna systems.Spatial Channel Correlation for MIMO system: Theoretically, the execution of wireless communication systems can be enhanced by having multiple antennas at the transmitter and the receiver. The thought is that if the proliferation channels between every match of transmit and get antennas are factually free and indistinguishably circulated, then multiple autonomous channels with indistinguishable attributes can be made by precoding and be utilized for either transmitting multiple information streams or expanding the unwavering quality (in terms of bit mistake rate).

9. Future Directions

The MIMO system gives distinctive increases, for example, exhibit pick up, diversity pick up, multiplexing addition and co-channel obstruction decrease. Every term which is previously mentioned has their own focal points, for example, cluster pick up builds scope and Quos(quality of administration) comparably the diversity pick up , multiplexing pick up increments ghastly effectiveness, and co-channel obstruction decrease increments cell limit. Most importantly when

OFDM technology it ended up being the best future technology for LTE systems.

10. Conclusion

In this paper there is fundamentally a study with respect to MIMO frameworks. This paper incorporates the investigation of MIMO alongside various generations from 1G to LTE frameworks, and then correspondence models are concentrated on beginning from SISO to SIMO to MISO lastly towards MIMO which are most broadly utilized as a part of various future advancements. In MIMO correspondence framework fundamentally their causes, rule, work and an alternate test confronted by this innovation is concentrated on.

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Professional articles:

1. Professional paper (contribution offering experience useful for improvement of professional practice but not necessarily based on scientific methods);
2. Informative contribution (editorial, commentary, etc.);
3. Review (of a book, software, case study, scientific event, etc.)

Language

The article should be in English. The grammar and style of the article should be of good quality. The systematized text should be without abbreviations (except standard ones). All measurements must be in SI units. The sequence of formulae is denoted in Arabic numerals in parentheses on the right-hand side.

Abstract and Summary

An abstract is a concise informative presentation of the article content for fast and accurate Evaluation of its relevance. It is both in the Editorial Office's and the author's best interest for an abstract to contain terms often used for indexing and article search. The abstract describes the purpose of the study and the methods, outlines the findings and state the conclusions. A 100- to 250-Word abstract should be placed between the title and the keywords with the body text to follow. Besides an abstract are advised to have a summary in English, at the end of the article, after the Reference list. The summary should be structured and long up to 1/10 of the article length (it is more extensive than the abstract).

Keywords

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Acknowledgements

The name and the number of the project or programmed within which the article was realized is given in a separate note at the bottom of the first page together with the name of the institution which financially supported the project or programmed.

Tables and Illustrations

All the captions should be in the original language as well as in English, together with the texts in illustrations if possible. Tables are typed in the same style as the text and are denoted by numerals at the top. Photographs and drawings, placed appropriately in the text, should be clear, precise and suitable for reproduction. Drawings should be created in Word or Corel.

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Footnotes are given at the bottom of the page with the text they refer to. They can contain less relevant details, additional explanations or used sources (e.g. scientific material, manuals). They cannot replace the cited literature.
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